

A Stable Layer Stripping Algorithm for Electrical Impedance Tomography

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EIT inverse problem

- Let $\Omega \subset \mathbb{R}^n$, $\sigma \in C^1(\Omega) \cap C(\bar{\Omega})$, and let $u \in H^2(\Omega)$ be the solution of

$$\begin{aligned} \nabla \cdot (\sigma \nabla u) &= 0 \text{ in } \Omega, \\ \sigma \frac{\partial u}{\partial n} \Big|_{\partial\Omega} &= f \in H_0^{1/2}(\partial\Omega), \end{aligned}$$

where

$$H_0^{1/2}(\partial\Omega) = \{f \in H^{1/2}(\partial\Omega) \mid \int_{\partial\Omega} f dS = 0\}.$$

- Neumann-to-Dirichlet map,

$$W = W[\sigma] : H^{1/2}(\partial\Omega) \rightarrow H^{3/2}(\partial\Omega), \quad \sigma \frac{\partial u}{\partial n} \Big|_{\partial\Omega} \mapsto u \Big|_{\partial\Omega}$$

- Calderón problem:** From the knowledge of W_σ , reconstruct σ

Layer Stripping and Invariant Imbedding

Background: One-dimensional (zero energy) Schrödinger equation:

$$-\psi''(x) + V(x)\psi(x) = 0.$$

Define the *wave impedance* $\eta(x)$,

$$\eta(x) = \frac{\psi'(x)}{\psi(x)}.$$

Differentiate:

$$\eta'(x) = \frac{\psi''(x)}{\psi(x)} - \left(\frac{\psi'(x)}{\psi(x)} \right)^2 = V(x) - \eta(x)^2.$$

Layer Stripping and Invariant Imbedding

- The wave impedance satisfies a Riccati equation

$$\eta'(x) = V(x) - \eta(x)^2.$$

- The wave impedance represents the Dirichlet-to-Neumann map:

$$\psi'|_{x=a} = \eta(a)\psi|_{x=a}.$$

- Moving the boundary $\{x = a\}$ is tantamount to solving the Riccati equation.
- The Neumann-to-Dirichlet map $\nu = 1/\eta$ (wave admittance) satisfies also a Riccati equation:

$$\nu'(x) = 1 - V(x)\nu(x)^2.$$

Layer Stripping and Invariant Imbedding

Extension to EIT: Let

$$D = \{(r, \theta) \mid 0 \leq r < 1\} \subset \mathbb{R}^2,$$

and u satisfies

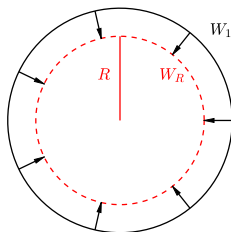
$$\begin{aligned} \nabla \cdot (\sigma \nabla u) &= 0 \text{ in } D, \\ \sigma \frac{\partial u}{\partial r} \Big|_{r=1} &= f. \end{aligned}$$

Neumann-to-Dirichlet map

$$W_1 : H^s(\partial D) \rightarrow H^{s+1}(\partial D), \quad f \mapsto u \Big|_{r=1},$$

where $s \geq -1/2$.

Layer Stripping and Invariant Imbedding



- 1 Define a family W_R of NtD operators over circles of radius R
- 2 Propagate the boundary data W_1
- 3 Evaluate the conductivity while marching in

Layer Stripping and Invariant Imbedding

1. Extension of the boundary data: Define

$$u_R(t, \theta) = u(Rt, \theta), \quad \sigma_R(t, \theta) = \sigma(Rt, \theta), \quad 0 < R \leq 1,$$

satisfying

$$\nabla_{\xi} \cdot (\sigma_R \nabla_{\xi} u_R) = 0 \text{ in } D,$$

where $\xi = (t \cos \theta, t \sin \theta)$.

$$W_R : H^{1/2}(\partial D) \rightarrow H^{3/2}(\partial D), \quad W_R \left[\sigma_R \frac{\partial u_R}{\partial t} \right]_{t=1} = u_R(1).$$

Layer Stripping and Invariant Imbedding

2. Propagation: Define

$$U = U(R, \theta) = \begin{bmatrix} v \\ w \end{bmatrix} \in H^{3/2}(\partial D) \times H^{1/2}(\partial D),$$

where

$$v = [u_R]_{t=1}, \quad w = \left[\sigma_R \frac{\partial u_R}{\partial t} \right]_{t=1}.$$

Differentiate with respect to R :

$$\dot{v} = \frac{\partial v}{\partial R} = \frac{1}{R\sigma(R, \theta)} \left[\sigma(Rt, \theta) \frac{\partial}{\partial t} u(tR, \theta) \right]_{t=1} = \frac{1}{R\sigma(R, \theta)} w,$$

or

$$\dot{v} = \frac{1}{R} Gw, \quad G : H^{1/2}(\partial\Omega) \rightarrow H^{1/2}(\partial\Omega).$$

Layer Stripping and Invariant Imbedding

$$\begin{aligned}
 \dot{w} &= \frac{\partial w}{\partial R} = \frac{\partial}{\partial R} \left[\sigma(R, \theta) R \frac{\partial u}{\partial R}(R, \theta) \right] \dot{w} \\
 &= -\frac{1}{R} \frac{\partial}{\partial \theta} \left[\sigma(R, \theta) \frac{\partial}{\partial \theta} u(R, \theta) \right] \\
 &= -\frac{1}{R} \frac{\partial}{\partial \theta} \left[\sigma(R, \theta) \frac{\partial}{\partial \theta} v \right],
 \end{aligned}$$

or

$$\dot{w} = -\frac{1}{R} S v, \quad S : H^{3/2}(\partial\Omega) \rightarrow H^{-1/2}(\partial\Omega).$$

Hence,

$$R\dot{U} = \begin{bmatrix} 0 & G \\ -S & 0 \end{bmatrix} U.$$

Layer Stripping and Invariant Imbedding

Differentiate

$$v = W_R w,$$

giving

$$\begin{aligned} \dot{W}_R w &= \dot{v} - W_R \dot{w} = \frac{1}{R}(Gw + W_R S v) \\ &= \frac{1}{R}(G + W_R S W_R) w \end{aligned}$$

Riccati equation:

$$R \dot{W}_R = G + W_R S W_R.$$

$$H_0^{1/2}(\partial D) \xrightarrow{W_R} H_0^{3/2}(\partial D) \xrightarrow{S_R} H_0^{-1/2}(\partial D) \xrightarrow{W_R} H_0^{1/2}(\partial D),$$

Layer Stripping and Invariant Imbedding

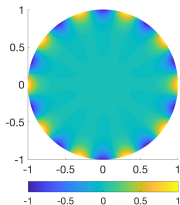
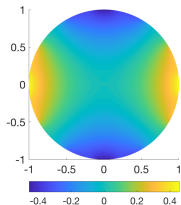
3. Reconstruction: High frequency asymptotics:

$$(W_R)_{jk} = \langle e^{ij\theta}, W_R e^{ik\theta} \rangle,$$

we have

$$\lim_{|k| \rightarrow \infty} |k| (W_R)_{n+k,k} = \frac{1}{2\pi R} \int e^{in\theta} \frac{1}{\sigma(R, \theta)} d\theta = \hat{\rho}_n,$$

where $\rho = 1/\sigma$ is the resistivity.



Layer Stripping

Layer stripping algorithm:

- Fix radii $1 = r_0 > r_1 > \dots > r_J > 0$. Denote

$$A_j = \{(r, \theta) \mid r_j < r < r_{j-1}\}, \quad 1 \leq j \leq J.$$

- Set $j = 0$
- For $j = 1 : J$ repeat
 - (a) Estimate $\sigma|_{A_j}$ from the high frequency components of $W_{r_{j-1}}$;
 - (b) Propagate $W_{r_{j-1}} \rightarrow W_{r_j}$ using the Riccati equation.

Somersalo E, Isaacson D, Cheney M and Isaacson E (1991) Layer stripping: A direct numerical method for impedance imaging. *Inverse Problems* **7** 899–926.

Cheney M, Isaacson D, Somersalo E and Isaacson E (1995) Layer stripping process for impedance imaging. U.S. Patent no. 5 390 110, February 14, 1995.

Layer Stripping

Ill-posedness: The ill-posed nature of the EIT inverse problem shows up as instability of the (backwards) Riccati equation.

Special case: Radial conductivity $\sigma = \sigma(r)$: The Fourier modes of W_R decouple. With $\sigma = 1$, we have

$$(W_R)_{kk} = w_k(R) = \frac{1}{|k|}.$$

The backwards Riccati initial value problem:

$$R \frac{dw_k}{dR} = 1 - k^2 w_k^2, \quad w_k(1) = b = \text{data}.$$

Layer Stripping

Explicit solution:

$$w_k(R) = \frac{1}{|k|} \frac{R^{2|k|} - M(|k|b)}{R^{2|k|} + M(|k|b)}, \quad M(|k|b) = \frac{1 - |k|b}{1 + |k|b}.$$

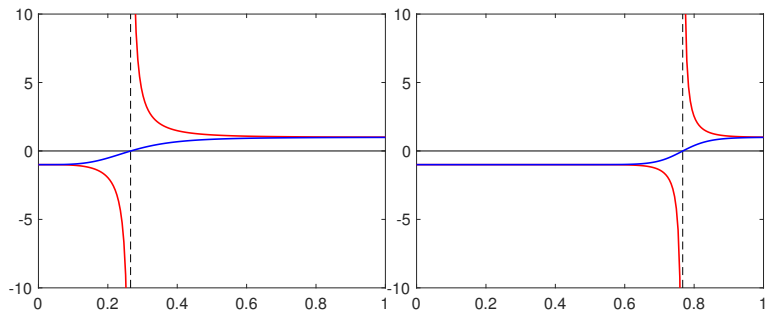
- Exact solution is recovered if the data are noiseless, that is, $b = 1/|k|$, implying that $M(|k|b) = 0$.
- Noisy data with relative error $\varepsilon \neq 0$,

$$b = \frac{1}{|k|} (1 \pm \varepsilon), \quad |\varepsilon| < 1.$$

The solution either becomes singular ($\varepsilon > 0$) or negative ($\varepsilon < 0$) at a radial value

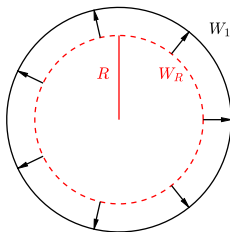
$$R \propto \frac{1}{|\varepsilon|^{1/2|k|}} \rightarrow 1, \text{ as } |k| \rightarrow \infty.$$

Layer Stripping



Solutions $|k|w_k(R)$ of the backwards Riccati equation, with positive (red) and negative (blue) error. $|\varepsilon| = 0.01$. On the left, $k = 2$, and on the right, $k = 10$. The correct solution is $|k|w_k(R) = 1$.

Reformulation



Forward problem in an annular domain: Given an $R < 1$, denote $A_R = \{(r, \theta) \mid R < r < 1\}$.

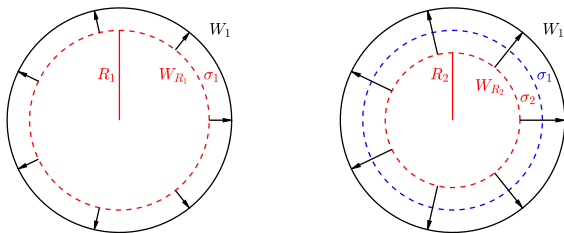
Consider the mapping

$$\Psi_R : (W_R, \sigma|_{A_R}) \mapsto W_1.$$

The forward problem is well-posed, as the forward Riccati propagation is stable.

Inverse problem: Given a noisy observation of W_1 , estimate $(W_R, \sigma|_{A_R})$.

Sequential estimation



- Sequence of radii $1 > R_1 > R_2 > \dots$
- Using W_1 , estimate (W_{R_1}, σ_1) , where $\sigma_1 = \sigma|_{A_{R_1}}$
- Using W_1 and the estimate for σ_1 , estimate (W_{R_2}, σ_2) , where $\sigma_2 = \sigma|_{A_{R_2}}$
- ...

Discretization

- Radii

$$1 = R_0 > R_1 > \dots > R_J > 0,$$

defining rings

$$A_j = \{(r, \theta) \mid R_j < r < R_{j-1}\}, \quad 1 \leq j \leq J.$$

- Approximation:

$$\sigma|_{A_j}(r, \theta) = \sigma_j(\theta), \quad R_j < r < R_{j-1}.$$

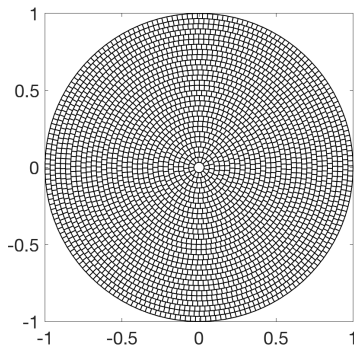
- Logarithmic parametrization of each σ_j :

$$(\lambda_j)_\ell = \log \frac{\sigma_j(\theta_\ell)}{\sigma_0}, \quad 1 \leq \ell \leq n_j, \quad \lambda_{J+1} = \log \frac{\sigma(0)}{\sigma_0},$$

where

$$n_j = \left\lfloor \frac{2\pi r_j}{h} \right\rfloor.$$

Discretization



Discretization

- Parametrization of the conductivity in $R_j \leq r \leq 1$:

$$\lambda_{(j)} = \begin{bmatrix} \lambda_1 \\ \vdots \\ \lambda_j \end{bmatrix} \in \mathbb{R}^{N_j}, \quad N_j = n_1 + \dots + n_j,$$

where $1 \leq j \leq J$.

- Interior boundary value:

$$w_j = \text{vec}(W_{R_j}) \quad (\text{stack the columns in a vector})$$

- Exterior boundary value: Numerical Riccati solver,

$$\psi_j(w_j, \lambda_{(j)}) = w_0.$$

Sequential estimation

State vectors, evolution model

$$x_0 \rightarrow x_1 \rightarrow x_2 \rightarrow \dots$$

Indirect observations,

$$b_j = F(x_j) + \varepsilon_j.$$

- 1 Given the probability distribution $\pi_k(x_k)$, propagate (push forward)

$$\pi_k(x_k) \rightarrow \tilde{\pi}_{k+1}(x_{k+1})$$

- 2 Using $\tilde{\pi}_{k+1}$ as prior, use Bayes' formula to update

$$\pi_{k+1}(x_{k+1}) \propto \tilde{\pi}_{k+1}(x_{k+1})\pi(b_{k+1} | x_{k+1})$$

- 3 Repeat

Sequential estimation

Define the *state vector*

$$x_k = \begin{bmatrix} w_0 \\ w_k \\ \lambda_{(k)} \end{bmatrix}.$$

For Bayesian filtering algorithm, we need

- 1 **State evolution model:** A stochastic model

$$x_k \rightarrow x_{k+1}.$$

- 2 **Observation model:** A stochastic model

$$x_k \rightarrow b_k.$$

Sequential estimation

1. State evolution model:

$$x_k = \begin{bmatrix} w_0 \\ w_k \\ \lambda_{(k)} \end{bmatrix} \xrightarrow{(a)} \begin{bmatrix} w_{k+1} \\ \lambda_{(k+1)} \end{bmatrix} \xrightarrow{(b)} \begin{bmatrix} w_0 \\ w_{k+1} \\ \lambda_{(k+1)} \end{bmatrix} = x_{k+1}.$$

- (a)** Given the current $\lambda_{(k)}$, draw λ_{k+1} from the conditional prior distribution $\pi_{\text{pr}}(\lambda_{k+1} \mid \lambda_{(k)})$ (smoothness prior for σ),
 Draw w_{k+1} from the prior distribution $\pi'_{\text{pr}}(w_{k+1})$.
- (b)** Propagate w_{k+1} through the $k + 1$ layers using the first order Möbius propagator.
 Add innovation with variance estimated from the second order Möbius propagator (numerical modeling error).

Sequential estimation

2. Observation model:

$$b_k = P x_k + \varepsilon_k,$$

where

$$P = \begin{bmatrix} I & O & O \end{bmatrix} : \begin{bmatrix} w_0 \\ w_k \\ \lambda^{(k)} \end{bmatrix} \mapsto w_0.$$

Gaussian observation error:

$$\varepsilon_k \sim \mathcal{N}(0, \Sigma).$$

Ensemble Kalman Filtering (EnKF)

- 1 **Propagation step:** Given a sample

$$\{x_k^{(1)}, x_k^{(2)}, \dots, x_k^{(N)}\}$$

from the current posterior, generate a predictive sample using the propagation model,

$$\{\hat{x}_{k+1}^{(1)}, \hat{x}_{k+1}^{(2)}, \dots, \hat{x}_{k+1}^{(N)}\}$$

- 2 Calculate the empirical mean \bar{x}_{k+1} and covariance G_{k+1} .
- 3 Given the observation b_{k+1} , generate a data ensemble

$$\{b_{k+1}^{(1)}, b_{k+1}^{(2)}, \dots, b_{k+1}^{(N)}\}, \quad b_{k+1}^{(j)} = b_{k+1} + e^{(j)}, \quad e^{(j)} \sim \mathcal{N}(0, \Sigma).$$

- 4 **Analysis step:** Generate a sample from the posterior by setting

$$x_{k+1}^{(j)} = \operatorname{argmin} \left\{ \|x - \hat{x}_{k+1}^{(j)}\|_{G_{k+1}}^2 + \|Px - b_{k+1}^{(j)}\|_{\Sigma}^2 \right\}.$$

Ensemble Kalman Filtering (EnKF)

Some observations:

- The forward model P is linear, and therefore the solution $x_{k+1}^{(j)}$ is obtained by a linear operation,

$$x_{k+1}^{(j)} = \widehat{x}_{k+1}^{(j)} + K(b_{k+1}^{(j)} - P\widehat{x}_{k+1}^{(j)}),$$

where K is the Kalman gain matrix

- The information about $b_{k+1}^{(j)}$ is passed to the parameter $\lambda_{(k+1)}$ through the cross covariance matrix

$$\text{cov}(w_0, \lambda_{(k+1)}).$$

- The algorithm, despite of nonlinearity, is **derivative-free**.

Likelihood model revisited

- The data b are formally used J times.
- A family of forward models,

$$b = \psi_j(\lambda_{(j)}, w_j) + \varepsilon_j, \quad \varepsilon_j \sim \mathcal{N}(0, C).$$

- Mean likelihood model (cf. Sequential Monte Carlo),

$$\begin{aligned} \pi_{\text{lkh}}(b \mid \lambda_{(J)}, w_{(J)}) &\propto \exp\left(-\frac{1}{2J} \sum_{j=1}^J \|b - \psi_j(\lambda_{(j)}, w_j)\|_C^2\right) \\ &\propto \prod_{j=1}^J \pi_{\text{lkh}}^j(b \mid \lambda_{(j)}, w_j), \end{aligned}$$

where

$$\pi_{\text{lkh}}^j(b \mid \lambda_{(j)}, w_j) \propto \exp\left(-\frac{1}{2J} \|b - \psi_j(\lambda_{(j)}, w_j)\|_C^2\right).$$

Posterior model

Bayes' theorem: (posterior \propto prior \times likelihood):

$$\pi_{\text{post}}(\mathbf{w}_{(J)}, \lambda_{(J)} \mid \mathbf{b}) \propto \pi_{\text{pr}}(\mathbf{w}_{(J)}, \lambda_{(J)}) \prod_{j=1}^J \pi_{\text{likh}}^j(\mathbf{b} \mid \lambda_{(j)}, \mathbf{w}_j).$$

Introduce the k th approximation of the posterior:

$$\pi_{\text{post}}^k(\mathbf{w}_{(k)}, \lambda_{(k)} \mid \mathbf{b}) \propto \pi_{\text{pr}}(\mathbf{w}_{(k)}, \lambda_{(k)}) \prod_{j=1}^k \pi_{\text{likh}}^j(\mathbf{b} \mid \lambda_{(j)}, \mathbf{w}_j).$$

Posterior model

- (a) A priori, w_{k+1} is independent of $w_{(k)}$ and $\lambda_{(k+1)}$.
- (b) A priori, λ_{k+1} is independent of $w_{(k)}$ but may not be independent of $\lambda_{(k)}$.

$$\pi_{\text{pr}}(w_{(k+1)}, \lambda_{(k+1)}) = \pi'(w_{k+1})\pi(\lambda_k | \lambda_{(k)})\pi_{\text{pr}}(w_{(k)}, \lambda_{(k)}),$$

Posterior model

Recursive updating:

$$\begin{aligned}
 \pi_{\text{post}}^{k+1}(\mathbf{w}_{(k+1)}, \lambda_{(k+1)} \mid \mathbf{b}) &\propto \pi_{\text{pr}}(\mathbf{w}_{(k+1)}, \lambda_{(k+1)}) \prod_{j=1}^{k+1} \pi_{\text{lkh}}^j(\mathbf{b} \mid \lambda_{(j)}, \mathbf{w}_j) \\
 &= \pi'_{\text{pr}}(\mathbf{w}_{k+1}) \pi_{\text{pr}}(\lambda_{k+1} \mid \lambda_{(k)}) \pi_{\text{pr}}(\mathbf{w}_{(k)}, \lambda_{(k)}) \prod_{j=1}^{k+1} \pi_{\text{lkh}}^j(\mathbf{b} \mid \lambda_{(j)}, \mathbf{w}_j) \\
 &= \pi'_{\text{pr}}(\mathbf{w}_{k+1}) \pi_{\text{pr}}(\lambda_{k+1} \mid \lambda_{(k)}) \pi_{\text{lkh}}^j(\mathbf{b} \mid \lambda_{(k+1)}, \mathbf{w}_{k+1}) \pi_{\text{post}}^k(\mathbf{w}_{(k)}, \lambda_{(k)} \mid \mathbf{b}) \\
 &= \underbrace{\left\{ \pi'_{\text{pr}}(\mathbf{w}_{k+1}) \pi_{\text{pr}}(\lambda_{k+1} \mid \lambda_{(k)}) \pi_{\text{post}}^k(\mathbf{w}_{(k)}, \lambda_{(k)} \mid \mathbf{b}) \right\}}_{(*)} \pi_{\text{lkh}}^j(\mathbf{b} \mid \lambda_{(k+1)}, \mathbf{w}_{k+1}).
 \end{aligned}$$

where (*) can be thought of as an updated prior for the next round.

Posterior model

Marginalize with respect to the initial values:

$$\bar{\pi}_{\text{post}}^k(\lambda_{(k)} | b) = \int \pi_{\text{post}}^k(w_{(k)}, \lambda_{(k)} | b) dw_{(k)}.$$

Integrating the recursive formula with respect to w_j :

$$\begin{aligned} \bar{\pi}_{\text{post}}^{k+1}(\lambda_{(k+1)} | b) &= \pi_{\text{pr}}(\lambda_{k+1} | \lambda_{(k)}) \bar{\pi}_{\text{post}}^k(\lambda_{(k)} | b) \\ &\times \left(\int \pi'_{\text{pr}}(w_{k+1}) \pi_{\text{lkh}}^j(b | \lambda_{(k+1)}, w_{k+1}) dw_{k+1} \right). \end{aligned}$$

Basis of the EnKF updating.

Solver for the Riccati equation

Action of $GL(2n)$ in the symplectic space \mathbb{R}^{2n} ,

$$z = \begin{bmatrix} z_1 \\ z_2 \end{bmatrix} \mapsto z' = Az = \begin{bmatrix} A_{11} & A_{12} \\ A_{21} & A_{22} \end{bmatrix} \begin{bmatrix} z_1 \\ z_2 \end{bmatrix}.$$

Assume that

$$z_1 = Wz_2.$$

We have

$$\begin{aligned} z_1' &= A_{11}z_1 + A_{12}z_2 = (A_{11}W + A_{12})z_2 \\ z_2' &= A_{21}z_1 + A_{22}z_2 = (A_{21}W + A_{22})z_2, \end{aligned}$$

implying that

$$z_1' = (A_{11}W + A_{12})(A_{21}W + A_{22})^{-1}z_2'.$$

Solver for the Riccati equation

Conclusion: The action of $A \in GL(2n)$ induces a transformation

$$W \mapsto W' = (A_{11}W + A_{12})(A_{21}W + A_{22})^{-1}$$

on the Grassmannian $\text{Gr}_n(2n)$.

Differential equations: Assume that $z = z(R) \in \mathbb{R}^{2n}$ satisfies

$$\dot{z} = Cz = \begin{bmatrix} C_{11} & C_{12} \\ C_{21} & C_{22} \end{bmatrix} \begin{bmatrix} z_1 \\ z_2 \end{bmatrix}, \quad z_1 = Wz_2.$$

Then,

$$\dot{z}_1 = \dot{W}z_2 + W\dot{z}_2,$$

or

$$\dot{W}z_2 = \dot{z}_1 - W\dot{z}_2.$$

Solver for the Riccati equation

Substitution:

$$\begin{aligned}
 \dot{W}z_2 &= \dot{z}_1 - W\dot{z}_2 \\
 &= C_{11}z_1 + C_{12}z_2 - W(C_{21}z_1 + C_{22}z_2) \\
 &= (C_{12} + C_{11}W - WC_{22} - WC_{21}W)z_2,
 \end{aligned}$$

that is, W satisfies the Riccati equation

$$\dot{W} = C_{12} + C_{11}W - WC_{22} - WC_{21}W.$$

Conclusion: A linear evolution model in \mathbb{R}^{2n} induces a Riccati a flow on the Grassmannian $\text{Gr}_n(2n)$.

Solver for the Riccati equation

First order scheme: Set

$$z(R+h) \approx z(R) + Cz(R)h = \underbrace{(I + hC)}_{=A} z(R).$$

Then, by the previous analysis, the first order propagation of W is obtained by

$$\begin{aligned} W(h) &\approx (A_{12} + A_{11}W)(A_{22} + A_{21}W)^{-1} \\ &= (hC_{12} + (I + hC_{11})W(R))(I + hC_{22} + hC_{21}W(R))^{-1} \end{aligned}$$

For the NtD problem, comparing the Riccati equations,

$$C = \frac{1}{R} \begin{bmatrix} 0 & G \\ -S & 0 \end{bmatrix}.$$

Solver for the Riccati equation

First order Möbius propagation scheme:

$$W(R+h) = \left(W(R) + \frac{h}{R} G(R) \right) \left(I - \frac{h}{R} S(R) W(R) \right)^{-1}.$$

Theorem

The eigenvalues of the matrix matrix $S(R)W(R)$ are all real and negative.

Forward propagation ($h > 0$) stable, while backwards propagation ($h < 0$) may encounter singularities.

Solver for the Riccati equation

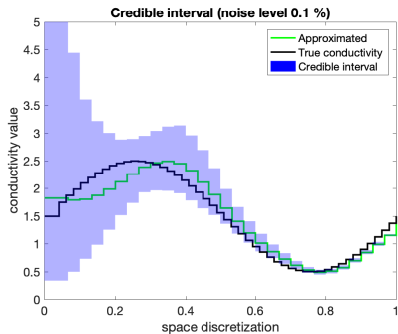
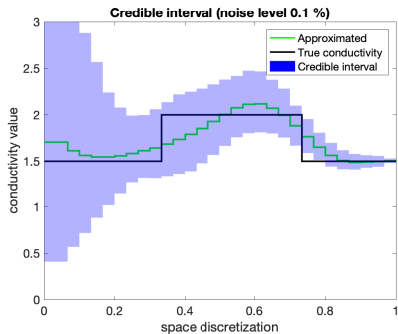
- The Grassmannian is a compact manifold, so the singularities of the Riccati equation are removable coordinate singularities.
- The Möbius solver, unlike standard RK or LMM solvers, have no problems going through the singularities.
- It is easy to define higher order solvers (error control for modeling approximation errors).

Computed examples

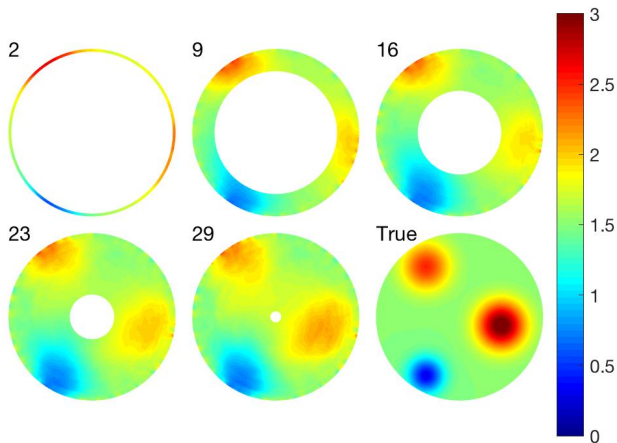
- EnKF with 1000 particles
- Number of frequencies = 60 (30 sines, 30 cosines)
- NtD map generated using a FEM approximation
- Additive white noise added
- Forward map: First order Möbius solver, accuracy controlled by using a second order Möbius solver.
- Radial case: Second order AR model to generate $\lambda_{j+1} \mid \lambda_{(j)}$.
- Non-Radial case: Use second order Gaussian smoothness prior, conditioning.

Computed examples

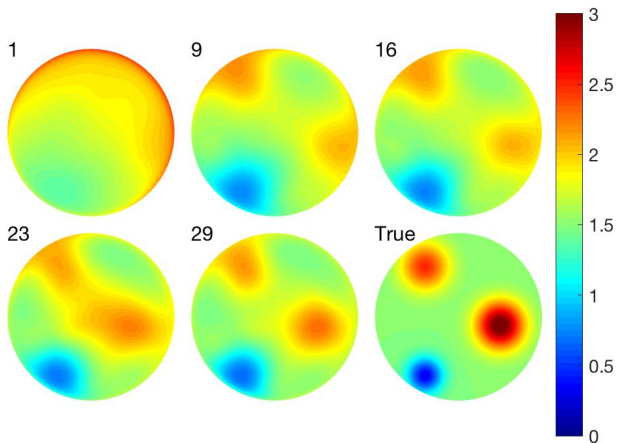
Radial conductivity



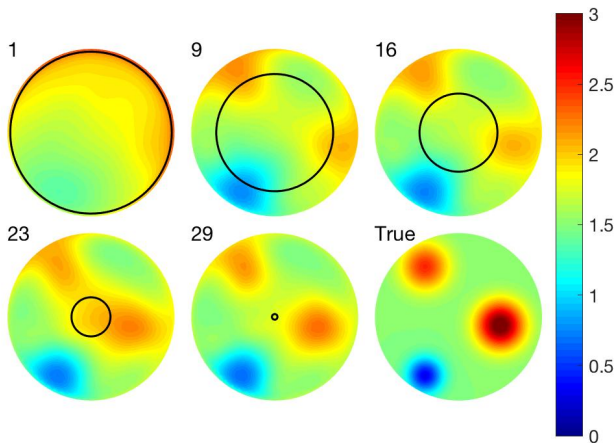
Results



Results



Results



Adding electrodes

- The layer stripping assumes that the data consist of the continuous Neumann-to-Dirichlet operator
- In reality, the EIT data is collected by using a finite number of contact electrodes
- Passing from electrode data to continuous data is in itself an ill-posed problem.

Complete Electrode Model (CEM)

$$\left\{ \begin{array}{ll} \nabla \cdot (\sigma \nabla v) = 0 & \text{in } \Omega, \\ \sigma \frac{\partial v}{\partial n} = 0 & \text{on } S_1 \setminus \cup e_\ell \text{ and } S_2 \\ v + z_\ell \sigma \frac{\partial v}{\partial n} = V_\ell & \text{on } e_\ell, 1 \leq \ell \leq L \\ \int_{e_\ell} \sigma \frac{\partial v}{\partial n} dS = J_\ell, & 1 \leq \ell \leq L, \end{array} \right. \quad (1)$$

Conservation of charge requires

$$\sum_{\ell=1}^L J_\ell = 0. \quad (2)$$

Data: Give $J \in \mathbb{R}^L$, measure $V \in \mathbb{R}^L$. Resistance matrix $R_\sigma : J \mapsto V$.

Connection with DtN

Theorem

Given $J \in \mathbb{R}_0^L$ and $f \in H^{1/2}(\partial\Omega)$, let $(v, V) \in \mathcal{H} = H^1(\Omega) \times \mathbb{R}_0^L$ be the solution of the CEM problem with applied current pattern J . Then

$$\int_{\partial\Omega} v \Lambda_\sigma f dS + \sum_{\ell=1}^L \frac{1}{z_\ell} \int_{e_\ell} (f - W_\ell)(v - V_\ell) dS - \sum_{\ell=1}^L J_\ell W_\ell = 0, \quad (3)$$

for all $W \in \mathbb{R}_0^L$.

Connection with DtN, Matrix form

Orthonormal basis in $H^{1/2}(\partial\Omega)$:

$$\varphi_0(\theta) = \frac{1}{\sqrt{2\pi}}, \quad \varphi_{2j}(\theta) = \frac{1}{\sqrt{\pi j}} \cos j\theta, \quad \varphi_{2j-1}(\theta) = \frac{1}{\sqrt{\pi j}} \sin j\theta, \quad j = 1, 2, \dots$$

Matrix representation of Λ_σ :

$$(\mathbf{L}_\sigma)_{jk} = \int_{\partial\Omega} \varphi_j \Lambda_\sigma \varphi_k dS = \langle \varphi_j, \Lambda_\sigma \varphi_k \rangle, \quad 0 \leq j, k < \infty,$$

where

$$\mathbf{L}_\sigma : \ell^2 \rightarrow \ell^2.$$

Connection with DtN, Matrix form

Orthonormal basis for current/voltage patterns:

$$(\Phi_m)_\ell = \sqrt{\frac{(2 - \delta_{m,L/2})}{L}} \cos \frac{2\pi}{L} m(\ell - 1), \quad 1 \leq \ell \leq L,$$

Representation of the resistance map in the basis Φ :

$$\tilde{R}_\sigma = \Phi^T R_\sigma \Phi \in \mathbb{R}^{(L-1) \times (L-1)},$$

Connection with DtN, Matrix form

Some auxiliary matrices:

$$D_{\ell\ell} = \frac{|e_\ell|}{z_\ell}, \quad 1 \leq \ell \leq L,$$

$$Y_{j\ell} = \frac{1}{|e_\ell|} \int_{e_\ell} \varphi_j dS, \quad 1 \leq \ell \leq L, \quad 0 \leq j < \infty,$$

$$M_{jk} = \sum_{\ell=1}^L \frac{1}{z_\ell} \int_{e_\ell} \varphi_j \varphi_k dS, \quad 0 \leq j, k < \infty.$$

Connection with DtN, Matrix form

Theorem

The matrices $L_\sigma : \ell^2 \rightarrow \ell^2$ and $R_\sigma \in \mathbb{R}^{L \times L}$ satisfy the identity

$$\Phi^T D \Phi - (Y D \Phi)^T (L_\sigma + M)^{-1} Y D \Phi = \tilde{R}_\sigma^{-1}, \quad (4)$$

where \tilde{R}_σ is the representation of the resistance map in the basis Φ ,

$$\tilde{R}_\sigma = \Phi^T R_\sigma \Phi \in \mathbb{R}^{(L-1) \times (L-1)},$$

- Computing R_σ from L_σ or its inverse is a well-posed problem
- The converse, estimating L_σ from R_σ is an ill-posed problem.
- For the stable layer stripping algorithm, only the stable form is necessary:

$$(\lambda_{(k)}, w_k) \mapsto w_0 \mapsto L_\sigma \mapsto R_\sigma.$$

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